Student paper

## A 2D-Non-Parabolic Six Moments Model

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The most accurate way to describe carrier transport is to solve the Boltzmann transport equation (BTE), for instance with the very time consuming Monte-Carlo (MC) technique. On an engineering level however macroscopic transport models are more efficient. Multiplication of the BTE with weight functions, approximation of the scattering integral with a macroscopic relaxation time and integration over k-space yields, for instance, the drift-diffusion model, the energy transport model, and the six moments model [1]. As demonstrated in [2] and [3] for the bulk case, a non-parabolic six moments model may be accurate down to 30 nm. The challenge is to model higher-order transport parameters like the energy relaxation time  $\tau_1$ , the second-order relaxation time  $\tau_2$ , the energy mobility  $\mu_1$  and the second-order mobility  $\mu_2$  (see Figure 2) with as few simplifying assumptions as possible. A good choice is the use of tabulated. data extracted from MC simulations [4]. So far, bulk data has been used to examine higher-order parameters in a device. However, important effects like surface roughness scattering or the quantization in inversion layers are not included in bulk MC-data.

In order to account for these effects, a selfconsistently coupled Subband-Monte-Carlo simulator (SMC) and a Schrödinger-Poisson (SP) solver (see Figure 1) are applied. The SP solves the quantum confinement and the SMC simulates the 2D transport in each subband [5]. In order to be consistent with the 2D-parameters the approach consistently, we have developed a 2D-higher-order-transport model. The following equations yield the 2D and 3D transport model with D as the dimension of the system. With  $D_0 = 1$ ,  $D_1 = 1$ ,  $D_2 = D/2$ ,  $D_3 = (2+D)/2$ ,  $D_4 = (2+D)D/2^{4-D}$ , and  $D_5 = (4+D)(2+D)/4$  the model reads

$$\phi_0: D_0 \partial_t n - \frac{1}{q_0} \nabla_r \mathbf{J_n} = 0,$$

$$\phi_2: D_2 \mathbf{k_B} \partial_t (nT_n) + \nabla_r \mathbf{S_n} - \mathbf{E} \cdot \mathbf{J_n} = -D_2 n \mathbf{k_B} \frac{T - T_0}{\tau_1},$$

$$\phi_4: D_4 \frac{\mathbf{k_B}^2}{m^*} \partial_t (nT_n^2 \beta) + \frac{2}{m^*} \nabla_r \mathbf{K_n} + \frac{4q_0}{m^*} \mathbf{E} \cdot \mathbf{S_n} = -D_4 \frac{n}{m^*} \mathbf{k_B}^2 \frac{T^2 - T_0^2}{\tau_2},$$

$$\phi_1: \mathbf{J_n} = \mu_0 H_0 (D_1 \nabla_r (\mathbf{k_B} nT_n) + h_0 q_0 \mathbf{E} n),$$

$$\phi_3: \mathbf{S_n} = -\mu_1 H_1 \left( D_3 \nabla_r \left( n \frac{(\mathbf{k_B} T_n)^2}{q} \beta \right) + h_1 D_3 \mathbf{k_B} T_n \mathbf{E} n \right),$$

$$\phi_5: \mathbf{K_n} = -\mu_2 H_2 \left( D_5 \nabla_r \left( n \frac{(\mathbf{k_B} T_n)^3}{q} \gamma \right) + h_2 D_5 (\mathbf{k_B} T_n)^2 \mathbf{E} n \beta \right).$$

Here,  $\phi$  denotes the moment of the equation. The even moments are the balance equations and the odd ones are the fluxes.  $H_i$  are the non-parabolicity factors. For parabolic bands,  $H_i = 1$ .  $\beta$  is the kurtosis and denotes the deviation from the Maxwellian distribution function (see Figure 4). For the 2D and the 3D case the kurtosis is defined as:

$$\beta = \frac{D}{D+2} n \frac{\langle \epsilon^2 \rangle}{\langle \epsilon \rangle^2}$$

In Figure 3 we show the output characteristics of the 2D drift-diffusion, energy transport, and the six moments model of a SOI-MOSFET.

We introduce a 2D-nonparabolic-six-moments model for scaled devices. Due to the parameter modeling based on MC-tables, important effects like quantization as well as surface roughness scattering are inherently included.

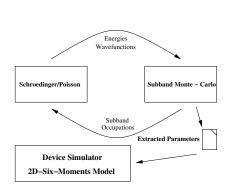


Figure 1: Transport parameters of a 2D-electron gas in an inversion layer are extracted selfconsistently and modeled through a whole device with a device simulator.

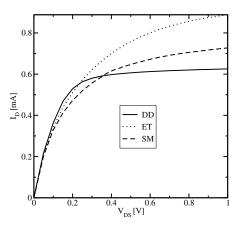


Figure 3: The output characteristic of a UTB SOI-Mosfet with a gate length of 40 nm calculated with the drift-diffusion model, the enery transport model, and the six moments model.

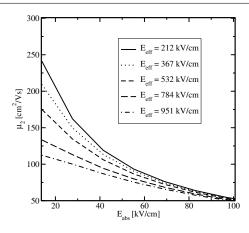


Figure 2: Extracted second-order mobility as a function of the driving field for different effective fields. The device simulator interpolates between these curves.

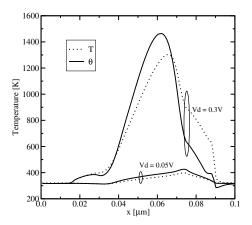


Figure 4: The second-order temperature  $\theta = \beta T$  in comparison to the carrier temperature T. With increasing drain voltage, the deviation from the Maxwellian distribution ( $\theta \approx T$ ,  $\beta \approx 1$ ) increases.

## References

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