

A Rigorous Study of Measurement Techniques for Negative Bias Temperature Instability

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Abstract—The active research conducted in the last couple of years demonstrates that negative bias temperature instability is one of the most serious reliability concerns for highly scaled pMOSFETs. As a fundamental prerequisite for a proper understanding of the phenomenon, accurate measurements are indispensable. Unfortunately, due to the nearly instantaneous relaxation of the degradation once the stressing conditions are removed, an accurate assessment of the real degradation is still extremely challenging. Consequently, rather than interrupting the stress in order to measure the degradation, alternative measurement techniques, such as the on-the-fly methods, which avoid stress interruption, have been proposed. However, these methods rely on rather simple compact models to translate the observed change in the linear drain current to a threshold voltage shift. As such, all methods have their own drawbacks which are rigorously assessed using a theoretical description of the problem.

Index Terms—Interface states, measurement, measure/stress/measure (MSM), negative bias temperature instability (NBTI), on-the-fly (OTF), oxide charges.

I. INTRODUCTION

WHEN SUBJECTED to negative bias temperature stress, device parameters of pMOS transistors have been observed to degrade [1]–[3]. This degradation is commonly described by a shift of the drain current or threshold voltage as a function of stress time. Due to the extremely fast recovery which sets in as soon as the stressing conditions are removed, any accurate assessment of the real degradation is very challenging.

In order to cope with that serious problem, many different measurement techniques have been proposed so far, which fall into different categories.

- 1) Conventional measure/stress/measure (MSM) techniques have been optimized by minimizing the delay between the removal of the stress and the first measurement point. In MSM techniques, either the drain current around the threshold voltage [4] is recorded and mapped to ΔV_{th} using an initial I_D – V_G curve around ΔV_{th} , or a threshold current is enforced through the device [5], thereby directly measuring V_{th} . Alternatively, complete I_D – V_G curves have been recorded using ultrashort pulses [6], [7]. It is still unclear whether it is

possible to reduce the delay sufficiently in order to record the true degradation characteristics [8], [9].

- 2) On-the-fly (OTF) techniques monitor the degradation of the drain current in the linear regime I_{Dlin} directly under stress conditions, thereby avoiding any relaxation. Different variants have been put forward [2], [10]–[12] which are more or less straightforward to implement [12].
- 3) In order to differentiate between oxide and interface charges, specialized measurement techniques such as charge-pumping (CP) and DCIV techniques have been used [3], [13]. However, due to the unclear interference of positive bias with the recovery characteristics, it is still very difficult to correctly interpret such data.

In order to link the observed degradation to possible underlying physical mechanisms, most commonly the creation of interface states and positive oxide charges [2], [3], a parametric relationship between the threshold-voltage shift/drain current and these charges is required. The natural candidate is the shift of the threshold voltage in the subthreshold regime, because it is directly related to the effective density of interface states and oxide charges via Poisson's equation as

$$\Delta V_{th}(t) = -\frac{\Delta Q_{it}(t) + \Delta Q_{ot}(t)}{C_{ox}}. \quad (1)$$

The positive charge stored in the interface states $Q_{it}(t)$ is given by $qN_{it}(t)f(V_G)$, where f denotes the hole occupancy of N_{it} depending on the gate (measurement) voltage V_G .

By contrast, I_{Dlin} as recorded during OTF techniques not only depends on Q_{it} and Q_{ot} , but also on the effective mobility μ_{eff} , which is known to change during degradation [1], [11], [14]. The three OTF extrapolation methods investigated here are based on a simple compact model (SPICE level-1)

$$I_D = \frac{\beta V_D (V_G - V_\theta - \frac{1}{2}V_D)}{1 + \theta (V_G - V_\theta - \frac{1}{2}V_D)}, \quad \text{for } V_G > V_\theta \quad (2)$$

or variants thereof, valid only in the linear regime under strong inversion [15].¹ The parameter β depends on the effective mobility, and θ describes the deviation from the linear behavior. Curiously, the fact that the threshold voltage V_θ used in expressions like (2) is the purely empirical *extrapolated threshold voltage* [12], [15] rather than the threshold voltage one obtains from the solution of the Poisson equation is rarely appreciated. This implies that changes in the extrapolated threshold voltage are not necessarily related in a 1:1 manner to additional charges

Manuscript received December 29, 2007; revised February 29, 2008. Current version published October 16, 2008.

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Digital Object Identifier 10.1109/TDMR.2008.2002353

¹We use the conventional nMOS sign convention in all formulas, with the sign change for pMOS devices being understood.

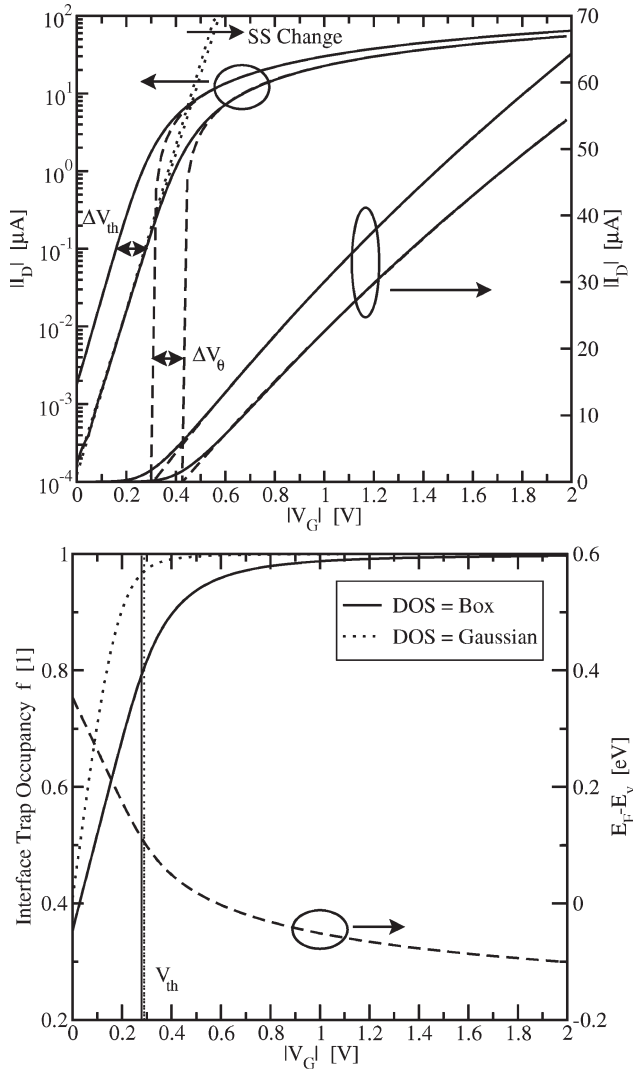


Fig. 1. **(Top)** Simulated I_D - V_G characteristics before and after 10^5 -s stress (shift to the right). It is assumed that, during stress, oxide and interface charges are accumulated ($N_{ot} = 1.3 \times 10^{12} \text{ cm}^{-2}$, $N_{it} = 7.2 \times 10^{11} \text{ cm}^{-2}$) which also lead to a 10% mobility degradation. Such an excessive degradation is, of course, unrealistic but results in a clearly visible impact and has been chosen for illustrational purposes only. It is clearly visible that the extracted ΔV_{th} and ΔV_{θ} are different quantities, with V_{θ} being larger than V_{th} [15]. Also indicated is the small change in the subthreshold slope caused by the changing interface state occupancy which affects the extraction of ΔV_{th} . **(Bottom)** Interface state occupancy f as a function of the gate voltage. During the measurement around the threshold voltage (indicated by thin straight lines), most traps are occupied (80%) for a flat DOS in the lower half of the bandgap. Also, shown is the gate voltage dependence of a Gaussian DOS with a maximum around 0.235 eV above the valence band edge [17], [18]. For a Gaussian DOS, a sharper transition, which occurs even later and gives an occupancy of about 97%, is observed. In order to consider the worst case, a (box) flat DOS is assumed in the following. Finally, the difference of the Fermi level E_F from the valence band edge is shown, resulting from Boltzmann statistic, which overestimate the hole concentration in strong inversion.

at the interface/oxide, also due to hidden correlations in the parameters V_{θ} , θ , and β [16].

In order to emphasize this issue, we use different symbols for V_{th} and V_{θ} . The difference of these two quantities is shown in Fig. 1. In particular, V_{θ} is always larger than V_{th} [15]. Furthermore, the exact value of V_{θ} obtainable from a calibration depends on the truncation of the measurement data used, because (2) is only valid for larger V_G . For instance, in

[19], only data with V_G larger than the value of V_G where the maximum slope is obtained is used, in order to guarantee a good fit of (2) in the strong inversion region. We remark that excellent fits of (2) are indeed possible in the strong inversion region [12], with increasing inaccuracies when V_G approaches V_{θ} .

One word of caution is in order: In the following study, we consider the threshold voltage shift in the subthreshold regime ΔV_{th} to be superior to ΔV_{θ} simply because it can be used to identify the buildup of interface and oxide charges via (1), provided possible distortions are small, because these charges are required for a correct physical understanding and modeling. Although from a circuit-designer point of view the shift in V_{θ} may be more relevant, it is more difficult to build a physics-based model on that data, as will be demonstrated in the following.

It is thus important to realize that none of the available measurement techniques is able to directly measure the real degradation; whereas MSM techniques have to cope with unavoidable delays, OTF techniques may also suffer from the conversion of ΔI_{Dlin} to ΔV_{th} , which is conventionally justified [2], [11] using rather simple compact models. As such, the results given by these techniques have been reported to differ by more than an order of magnitude in extreme cases [3], [11]. Although this is commonly attributed to the delay inherent in MSM techniques [3], we will show that this issue is more involved by conducting a thorough theoretical study using a numerical device simulator enhanced by appropriate reliability models.

II. PHYSICAL MODELING

A comparison of MSM and OTF results is shown in Fig. 2 for a typical IMEC device [4]: While the MSM technique can estimate the true degradation only via extrapolation (we use our technique [9], [20] based on the universality of negative bias temperature instability (NBTI) degradation [20], [21]), the ΔV_{th} values predicted by the simplest OTF method do not fit the detailed relaxation data very well, regardless of any extrapolation uncertainties.

In order to study this discrepancy, we turn the problem around: We place a certain predefined amount of ΔQ_{it} and ΔQ_{ot} at the interface of a pMOS and numerically simulate the drain current observed in the various techniques. For our simulations, we employ a suitably extended version of our numerical device simulator MINIMOS-NT [22]. Carrier transport is described using the drift-diffusion formalism which delivers accurate drain currents for small V_D and not too small channel lengths [23]. Boltzmann statistics are assumed for the carrier concentrations, resulting in an overestimation for large gate overdrive. The impact of these assumptions to the assessment of OTF data needs to be considered in greater depth, although we do not expect the inclusion of Fermi-Dirac statistics to significantly alter our results. We include accurate models for the interface state dynamics (SRH) with a realistic density-of-state (DOS) distribution [20], [24] and the mobility variation due to Coulomb scattering induced by interface states [25]. For pure silicon dioxide, interface states have often been assumed to introduce donorlike Gaussian peaks close to the valence band edge (we use 0.235 eV [17], [18]) into the DOS. For the device

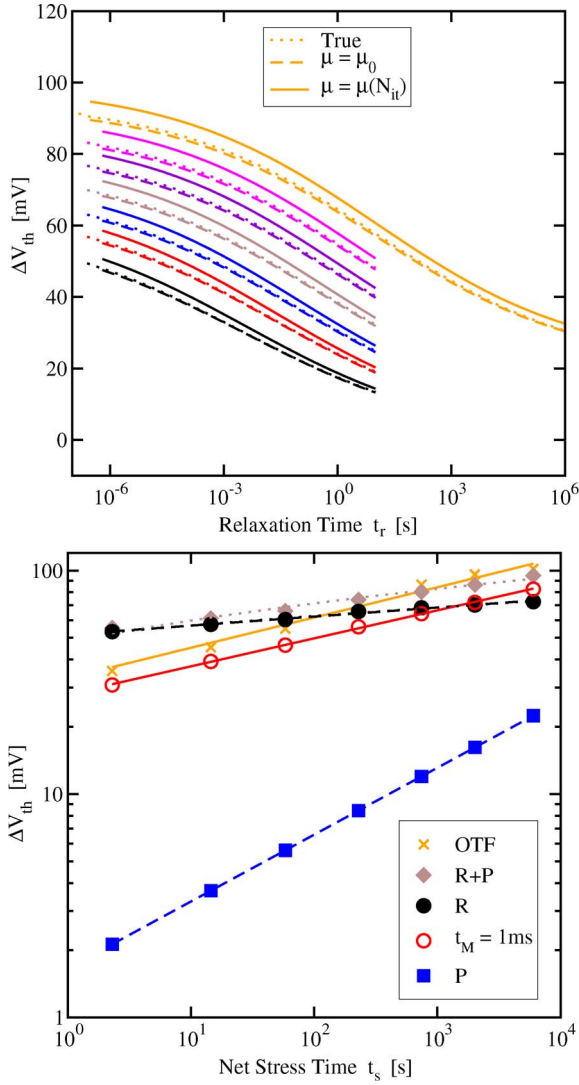


Fig. 2. Comparison of (open symbols) measured MSM and OTF1 data. **(Top)** Recorded detailed relaxation sequence allows one to reconstruct the (closed symbols and lines) full NBTI degradation by identifying a universally relaxing component R which is observed on top of a permanent/slowly relaxing component P . The degradation observed in OTF1 measurements does not agree well with the relaxation data. **(Bottom)** Extracted components R and P , the OTF1 data, and the MSM data with a delay of $t_M = 1$ ms as a function of the stress time. Clearly, the overall degradation $R + P$ is only in poor agreement with the OTF1 data.

under consideration, this implies that, around the threshold voltage, practically all interface states are occupied by holes. Recently, it has been pointed out that the DOS in oxynitrides may be markedly different [24]. Because the occupancy of the interface states determines the “visible” charge in the interface states, a changing occupancy during I_D – V_G measurements changes the subthreshold slope and may consequently distort the extracted V_{th} . In order to consider the worst case, we use a flat DOS in the lower half of the bandgap in the following, the impact of which is compared to a Gaussian distribution in Fig. 1. Furthermore, all simulations are based on a numerical device description carefully calibrated to measured I_D – V_G curves for a pMOS with $L_g = 0.5 \mu\text{m}$ and EOT = 1.4 nm.

The simulated drain current is then processed analogously to the measurement data and converted into ΔV_{th} . We then

compare whether the true ΔV_{th} given by the predefined charges agrees with what is observed in the simulated measurements. Consequently, this approach allows us to study the sensitivity of the extracted ΔV_{th} to the measurement setup and possible distortions.

Lacking a valid NBTI model that covers the degradation during both stress and recovery, as needed for simulated measurements, we rely on our empirical model in the following [9]: A growing number of recent publications suggests [2], [3], [26] that the degradation caused by NBTI consists of two components, a fast universally recoverable component R on top of a slowly recovering or permanent component P [8], [9]. The mapping of these two components on oxide or interface charges is still under debate, but it is often suggested that the recoverable component R is due to oxide charging and discharging, whereas P is caused by interface states [2]. Alternatively, it has been suggested that NBTI is primarily determined by interface states [3].

We empirically describe these two components as

$$R(t_s, t_r) = A \log \left(1 + \frac{t_s}{\tau_R} \right) r(t_r/t_s) \quad (3)$$

$$P(t_s) = \frac{P_{\max}}{1 + (t_s/\tau_P)^{-\alpha}} \quad (4)$$

with the universal relaxation function

$$r(\xi) = \frac{1}{1 + B\xi^{\beta_r}} \quad (5)$$

and the parameter values $A \approx 2.7$ mV, $\tau_R \approx 1.4$ ns, $B \approx 3.2$, $\beta_r \approx 0.16$, $P_{\max} \approx 50$ mV, $\tau_P \approx 13$ ks, and $\alpha \approx 0.45$. We remark that these values are determined for relaxation at $V_G \approx V_{th}$ and vary for other bias conditions. For the extraction of the parameters, the algorithm detailed in [8] and [9] has been applied to the data shown in Fig. 2. Consequently, our empirical description is accurate only under this condition and does not describe relaxation at varying V_G (as would be needed for a correct analysis of I_D – V_G and CP data). In the following, we will consider the two cases:

- 1) The degradation is due to oxide and interface states, assuming $\Delta Q_{ot}(t_s, t_r) = R(t_s, t_r)$ and $\Delta N_{it}(t_s) = P(t_s)$.
- 2) The degradation is due to interface states only, $\Delta N_{it} = R(t_s, t_r) + P(t_s)$.

The bias dependence of the oxide charges is hardwired in the universal relaxation function for a recovery around $V_G \approx V_{th}$, whereas the actual charge stored in the interface states depends on the position of the Fermi level (and thus the gate voltage) via the density-of-states. For our choice of parameters [24], we found that during stress all interface states are positively charged, whereas during relaxation (measurement around V_{th}), the occupancy goes down to $\sim 80\%$ (see Fig. 1).

For the sake of completeness, we remark that the widely used reaction-diffusion model [1], [3], given by $R = At_s^n r(t_r/t_s)$, $n = 1/6$, $B = 1$, and $\beta_r = 1/2$, is unable to capture any relaxation data known to us [9], [20].

In the following, the measurement methods discussed in this paper are summarized. We also highlight potential distortions due to often unappreciated physical effects like mobility variations and interface state dynamics.

A. MSM Methods

First, we use the method suggested in [4] which records several relaxation phases after exponentially growing stress intervals. As such, a maximum amount of information on a single device in a single measurement cycle is gathered. The recorded drain current around V_{th} is then converted into ΔV_{th} using an initial (narrow) I_D – V_G curve. Using the method developed in [8], a fast component R on top of a permanent component P is extracted and mapped to ΔQ_{ot} and ΔQ_{it} using the previously stated prescriptions.

In addition to this $I_D @ V_{th}$ method, we calculate fast (1 μ s) I_D – V_G curves for the extraction of ΔV_{th} using two constant current criteria. We remark that the relaxation predicted by our model is accurate only when switching the bias from the stress level to $V_G = V_{th}$, and the estimates obtained using this method must be taken with care.

B. OTF Methods

The first and simplest OTF method we discuss (OTF1) assumes θ and V_D to be small and also neglects any change in the effective mobility μ_{eff} (and thus β) [1], [3]

$$\Delta V_{\theta}^{(1)} \approx \frac{I_D - I_{D0}}{I_{D0}} (V_G - V_{\theta 0}). \quad (6)$$

The OTF1 method has the advantage that it only requires the determination of the initial drain current in the linear regime under stress conditions. The neglect of the mobility change (β and θ) in the OTF1 method has motivated the proposal of the OTF2 method which is capable of monitoring a mobility variation by introducing small perturbations of the stress voltage [2]

$$\Delta V_{\theta}^{(2)} \approx \sum_n \frac{\frac{I_D[n]}{g_m[n]} - \frac{I_D[n-1]}{g_m[n-1]}}{1 + 2\theta (V_G - V_{\theta}[n] - \frac{1}{2}V_D)}. \quad (7)$$

However, OTF2 still neglects the temporal change of θ . Furthermore, the OTF2 technique requires the determination of a full initial I_D – V_G curve for the determination of θ . Thus, as an alternative, one may use a variant suggested by Zhang and Chang [11] (OTF3)

$$\Delta V_{\theta}^{(3)} \approx \sum_n \frac{I_D[n] - I_D[n-1]}{\frac{1}{2}(g_m[n] + g_m[n-1])}. \quad (8)$$

III. POSSIBLE DISTORTIONS

The analysis of measurement data often assumes that the change in the drain current is only caused by ΔV_{θ} due to oxide/interface charges. However, it has been pointed out [2], [11], [14], [27] that mobility variations may impact the accuracy of the methods. This issue is commonly associated with

OTF techniques, but a rigorous theoretical analysis has not been conducted so far. This will be done in the following, where we also demonstrate the impact of mobility variations on MSM techniques. In addition, the response of the interface states on rapidly changing gate voltages is discussed.

A. Mobility Variations

1) *OTF Methods:* To analyze the sensitivity of the extracted ΔV_{θ} to changes in the mobility, we estimate the drain current in the linear region by a further simplified version of (2) ($\theta \approx 0$, $V_D \ll V_G$)

$$I_D = \beta V_D (V_G - V_{\theta}). \quad (9)$$

According to OTF1, the change in the linear drain current is converted to a threshold-voltage shift using $\Delta V_{\theta} = \Delta I_D / \beta$. Because β is approximately linearly proportional to μ_{eff} , according to (9), a change in $\mu_{eff} = \mu_{eff0} + \Delta\mu_{eff}$ results in

$$\Delta I_D = I_{D0} \Delta\mu_{eff} / \mu_{eff0}. \quad (10)$$

Assuming that this shift ΔI_D is *only* caused by mobility variations, a spurious shift in ΔV_{θ} is obtained as

$$\Delta V_{\theta}^{\mu} = -\frac{\Delta\mu_{eff}}{\mu_{eff0}} (V_G - V_{\theta}) = -(V_G - V_{\theta}) \frac{p}{100} \quad (11)$$

with p being the mobility variation in percent. With $V_G = 2$ V and $V_{\theta} \approx 0.25$ V (fit to the linear part of the on-current), we obtain that the error in the extracted ΔV_{θ} equals 17.5 mV for each percent mobility error, rather much larger than previously anticipated [1]. As an example, for an error in the effective mobility of 3% [1], one obtains an error of about 50 mV in ΔV_{θ} , which is considerable keeping in mind that a 100 mV shift is often considered to cause failure.

2) *MSM Methods:* By contrast, data obtained via the MSM technique is much less sensitive to mobility changes. This can be easily demonstrated starting from a simple compact model for the subthreshold current [15]

$$I_D = A \mu_{eff} \exp\left(\frac{V_G - V_{th}}{\eta V_T}\right). \quad (12)$$

Here, V_T is the thermal voltage $k_B T / q$. Although the current in the subthreshold regime is also proportional to the mobility change and, consequently, a change in the mobility results also in an equivalent change of I_D via (10), the drain current depends now *exponentially* on the threshold-voltage shift. Thus, by inverting (12), one obtains

$$\Delta V_{th}^{\mu} = -\eta V_T \log\left(1 + \frac{\Delta\mu_{eff}}{\mu_{eff0}}\right) \approx -\eta V_T \frac{\Delta\mu_{eff}}{\mu_{eff0}} \quad (13)$$

$$= -\eta V_T \frac{p}{100}. \quad (14)$$

For the device under consideration, $\eta \approx 1.14$ (subthreshold slope of about 68 mV/dec at room temperature), giving $\eta V_T \approx 42$ mV and, finally, a 420- μ V error for each percent variation in the mobility. However, because the variation of the low-field mobility is expected to be larger than the high-field mobility [1], [11], for instance, 10%, one obtains an error in ΔV_{th} of about

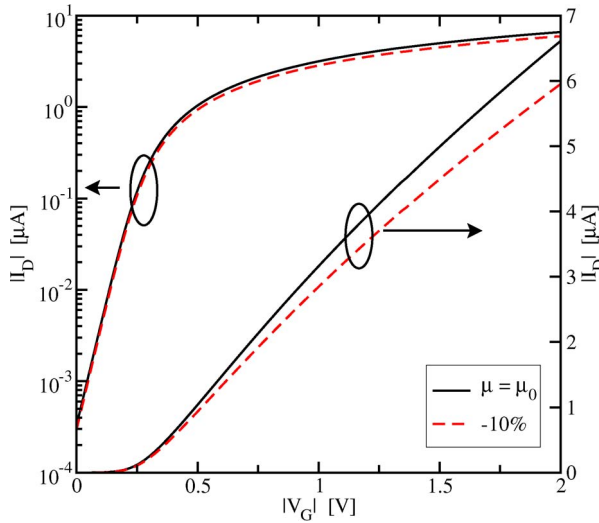


Fig. 3. Simulated I_D - V_G characteristics at $V_D = -50$ mV with nominal mobility values and an excessively large mobility degradation of about 10% for illustrative purposes. Clearly, the impact is largest in the linear regime.

5 mV, which is about a factor of ten lower than the corresponding error in the OTF method. Interestingly, the mobility-variation-induced error of the MSM method depends linearly on temperature via V_T , whereas the error of the OTF has only a weak temperature dependence via V_{th} .

3) *General Gate-Voltage Dependence for Mobility Variations*: The previous two results clearly show that the impact of the mobility variation on the estimated threshold-voltage shift depends on the operation mode of the transistor and, thus, on the applied gate voltage. Fig. 3 shows the change in the numerically simulated I_D - V_G characteristics caused by a 10% mobility degradation. Clearly, the impact is largest in the linear regime. The spurious threshold-voltage shift caused by this mobility variation is obtained by determining the difference in V_G at the same current I_D . For reasonable changes in the mobility, this spurious ΔV_{th}^μ depends linearly on the mobility variation. Fig. 4 shows ΔV_{th}^μ per percent mobility variation p as a function of gate bias. The smallest dependence of $\approx 500 \mu V \times p$ is obtained in the subthreshold regime (in that example, $V_G < 0.2$ V), where I_D perfectly follows (12). This value is in good agreement with the analytic estimate given by (14). Consequently, in order to avoid mobility-induced distortions of ΔV_{th} , measurements should be conducted with V_G safely in the exponential regime of I_D . For larger V_G , on the other hand, ΔV_{th}^μ rapidly increases in accordance with (11), which is also in good agreement with the full numerical result.

B. Dynamic Response of the Interface State Occupancy

An issue particularly relevant for ultrafast MSM measurements is the response of the interface states to a quickly changing gate voltage. It has been argued that, for measurements in the microsecond regime, the interface states may not react fast enough to the rapid drop in V_G and thus distort the response of I_D and, consequently, the estimated ΔV_{th} [27], [28]. From a theoretical point of view, the dynamics are governed by dynamic SRH kinetics, which is straightforward to include in a

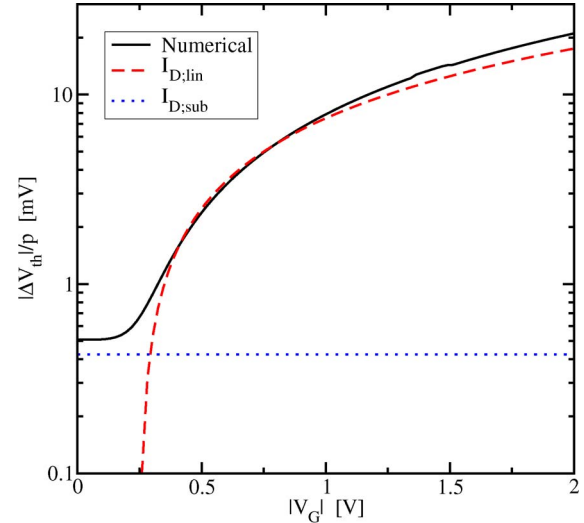


Fig. 4. Spurious threshold-voltage shift ΔV_{th}^μ per percent change in μ_{eff} obtained from Fig. 3 by considering the difference in V_G at the same I_D . With increasing V_G , ΔV_{th}^μ increases. Also, shown are the error estimates for the (dashed lines) linear regime (11) and the (dotted lines) subthreshold regime (14).

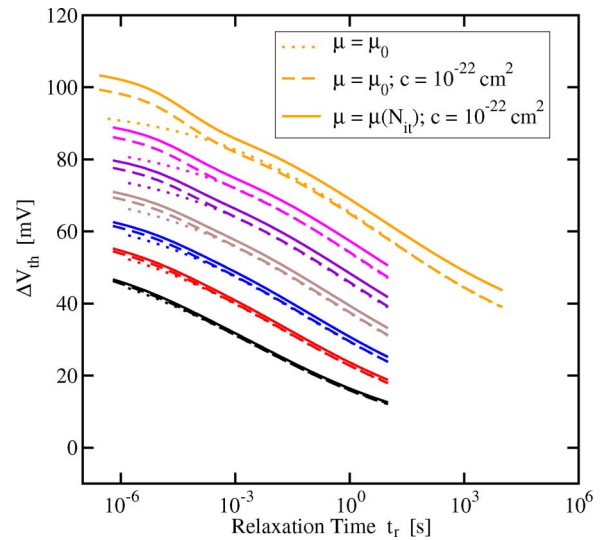


Fig. 5. Simulated impact of a 10% mobility variation and the dynamics of the interface state occupancy on the extracted ΔV_{th} . The same stress times as in Fig. 2 are used. Because the mobility model used depends only on N_{it} , which is assumed to be permanent, the mobility variation causes a constant offset in each relaxation curve (4.5 mV in the last curve at $t_s = 6000$ s). Also, only for unusually small capture cross sections c on the order of 10^{-22} cm^2 , initial distortions of ΔV_{th} can be observed.

numerical simulation. Again, a flat DOS is used for the interface states in order to maximize the impact of this effect (worst-case estimation). The timescales are determined by the capture cross sections, somewhat elusive quantities [29], and by how far the Fermi level is moved away from the valence band edge. In order to accurately reproduce CP measurements, capture cross sections on the order of 10^{-16} cm^2 are used in our simulator. Note, however, that values in the range $10^{-14}, \dots, 10^{-18} \text{ cm}^2$ have been reported [29], [30]. Interestingly, an impact of interface state dynamics on the simulated I_D in the microsecond regime is only observed for unusually small capture cross sections on the order of 10^{-22} cm^2 (cf. Fig. 5), in disagreement with the

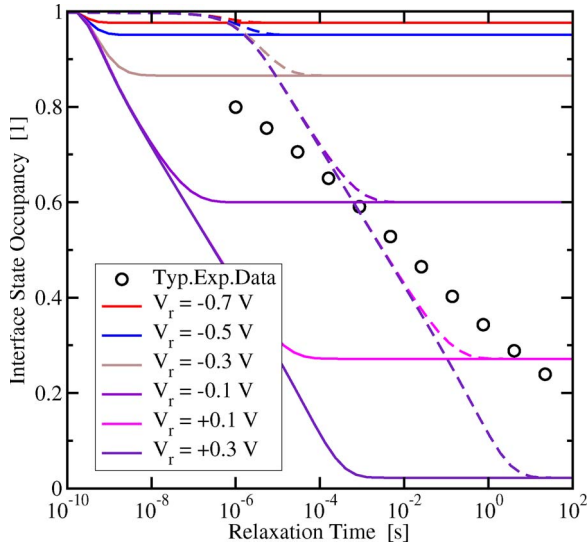


Fig. 6. Simulated response of the interface state occupancy when V_G is switched to a smaller recovery voltage. The dashed lines are obtained with extremely small capture cross sections of 10^{-22} cm^2 .

assertions made in [27]. Such small capture cross sections can only be justified by assuming defects at a certain distance from the interface which require tunneling for charge exchange [31]. We also remark that such a process, if occurring on top of the fast component R , would result in a visible distortion of the relaxation of I_D , not present in any of the data available to us [9]. Thus, we prefer the notion that if such a process is relevant at all, it should naturally be part of the fast component R .

To understand the dynamic response more clearly, it is intuitive to investigate the step response of the interface states when V_G is switched from stress bias (-2 V) to various recovery biases (cf. Fig. 6). Although the response is “loglike” [27], it is too fast to be significant for negative recovery voltages and conventionally used capture cross sections.

IV. RESULTS

In the following, we introduce the analytic models for Q_{ot} and Q_{it} into the device simulator, resimulate the various measurement sequences, and compare the extracted with the expected values.

A. MSM Methods

We begin with the MSM method used to originally extract the temporal evolution of Q_{ot} and Q_{it} . Quite convincingly, the simulation reproduces the relaxation data at $V_G \approx V_{th}$ and captures the trend in the OTF1 data at $V_G = -2 \text{ V}$ (cf. Fig. 7). The extracted ΔV_{th} with zero delay (frozen relaxation) and neglected mobility changes agrees perfectly with the expected values given through Q_{ot} and Q_{it} . However, when the degradation of the mobility is considered (about 10% at 10^5 s [1]), an error of about 4 mV at $t_s = 6000 \text{ s}$ is observed. Because the mobility model used assumes only the permanent interface states to affect the mobility, no relaxation in this error is observed. Consequently, this error translates directly to the permanent component. In reality, oxide charges located farther

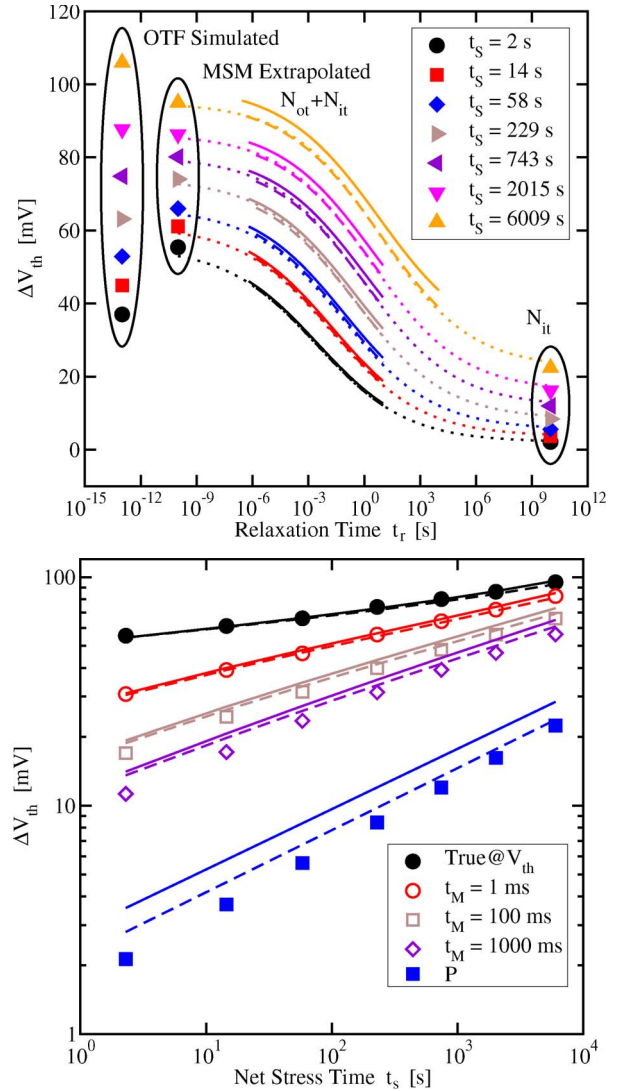


Fig. 7. Simulation of the measurement sequence shown in Fig. 2, which exactly reproduces the measurement results used to extract N_{it} and N_{ot} without any serious distortion. (**Top**) Extracted ΔV_{th} shift, (solid and dashed lines) with and without considering the mobility variation, confirming that the extraction is only weakly dependent on the mobility variation. The dotted lines are a guide to the eye taken from Fig. 2. (**Bottom**) Same simulation results as a function of the stress time. The (lines) simulation results agree very well with the (open symbols) measurement data and the (closed symbols) extracted P and R components. A 4-mV error is introduced in the permanent component (N_{it}) when the 10% mobility degradation is considered.

away from the interface also will affect the mobility; therefore, a slight change in this result is to be expected.

In addition to confirming the good accuracy of the MSM method, the influence of the measurement delay is exactly reproduced. In particular, although strongly dependent on the measurement delay, the extracted ΔV_{th} is practically independent of the method used (conversion of $I_D(V_{th})$, see [4], various constant I_D criteria, ultrafast pulses, etc.) and the stress-dependent change in the mobility (cf. Fig. 8).

B. OTF Methods

A more involved behavior is observed for ΔV_{th} extracted using the three OTF methods under consideration (cf. Fig. 9).

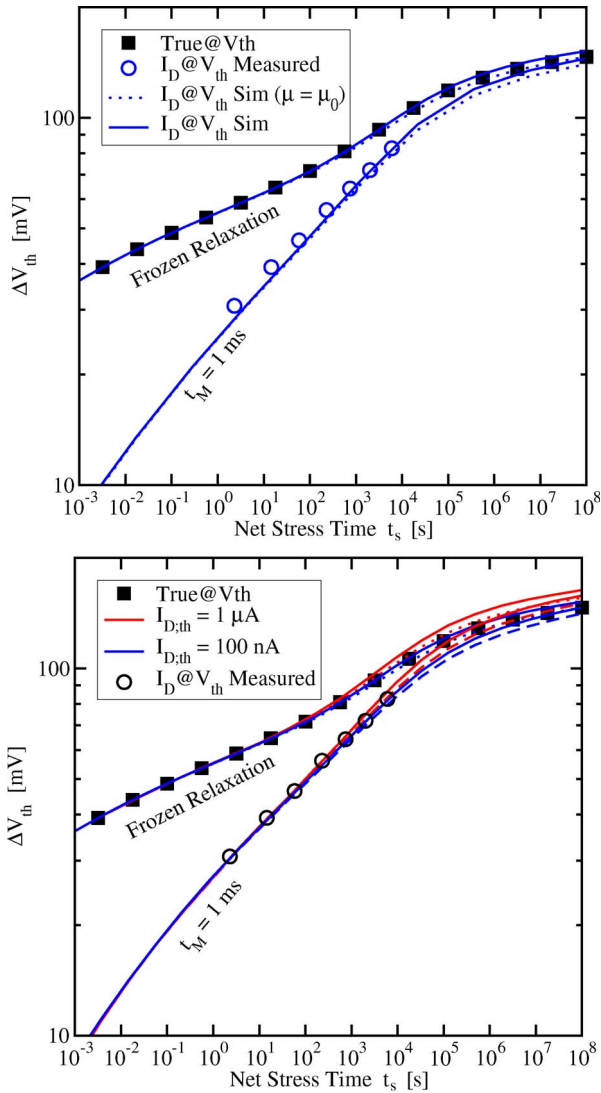


Fig. 8. Ideal no-delay MSM measurement exactly reproduces the true ΔV_{th} . In particular, (solid versus dotted lines) the extracted ΔV_{th} is only weakly dependent on the mobility change and the extraction method. Shown are extractions (top) from I_D at V_{th} and (bottom) from $I_D - V_G$ curves with various $I_{D,th}$ values. With increasing $I_{D,th}$, a small additional error is visible due to the increased occupancy of N_{it} .

All OTF methods suffer from the delay inherent in the first measurement point [32] which strongly distorts the data and is the equivalent problem of the delay in MSM techniques. We remark, however, that an exact prediction of the influence of the initial delay is difficult and may be exaggerated by assuming the universal relaxation law to be valid also for very short times.

In addition, the three techniques produce markedly different results, with only the OTF2 method [2] being insensitive to mobility changes (cf. Fig. 10). Note also that even extremely fast determination of the initial value I_{D0} leaves a residual error in the OTF methods due to the use of compact models. This is a rather severe disadvantage considering the widespread use of the OTF1 method in particular [1], [3].

The deficiencies of the OTF methods are also shown in Fig. 11 where the extracted slopes for OTF1 and OTF3 show a considerable error due to the mobility change, in contrast

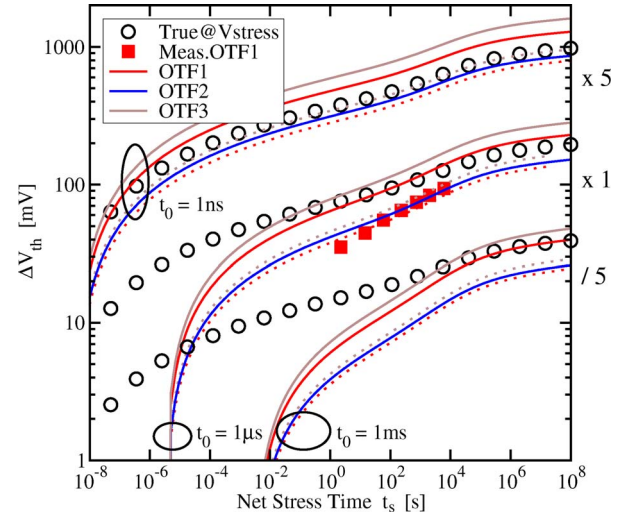


Fig. 9. Influence of the initial delay t_0 and the mobility change on OTF results. Whereas all OTF methods are strongly affected by the delay of the initial measurement (data scaled by a factor of five for readability), only OTF2 is not affected when (solid lines) the full interface charge dependent mobility model is used.

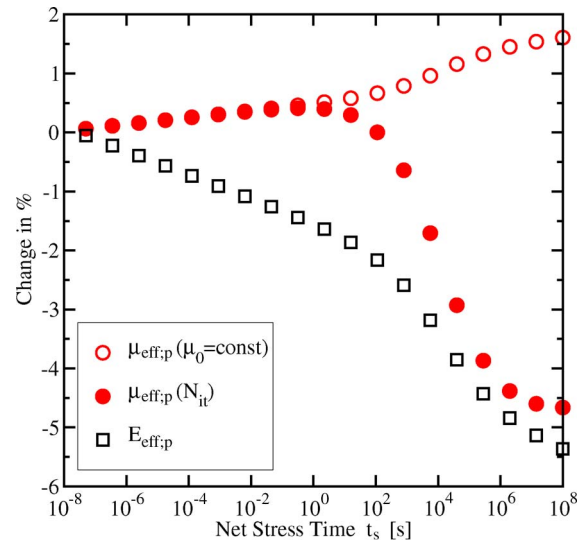


Fig. 10. Time dependence of the effective mobility ($\int \mu_p p dy / \int p dy$) and the effective electric field ($\int E_y p dy / \int p dy$) during the simulated measurement. Here, p is the hole concentration in the channel. When the influence of the interface charges is turned off in the mobility model, the effective mobility increases due to the reduction of $|V_G - V_{th}|$, (see [11]), whereas the inclusion of the additional Coulomb scattering at the interface leads to a degradation of about 4% at $t_s = 10^6$ s (see [1]).

to the MSM method. In addition, we wish to remark that the theoretical accuracy obtainable by the OTF2 method is difficult to realize in practice, because the extracted ΔV_{th} is extremely sensitive to unavoidable measurement errors in the estimation of g_m [12].

Finally, we check the potential errors when R and P are assumed to be due to interface states only. Fig. 12 shows the impact of a 10% mobility variation on the extracted ΔV_{th} . In contrast to the first case where oxide and interface charges were assumed, the mobility degradation causes a constant 5% error in the extracted data. Similarly, for the OTF methods,

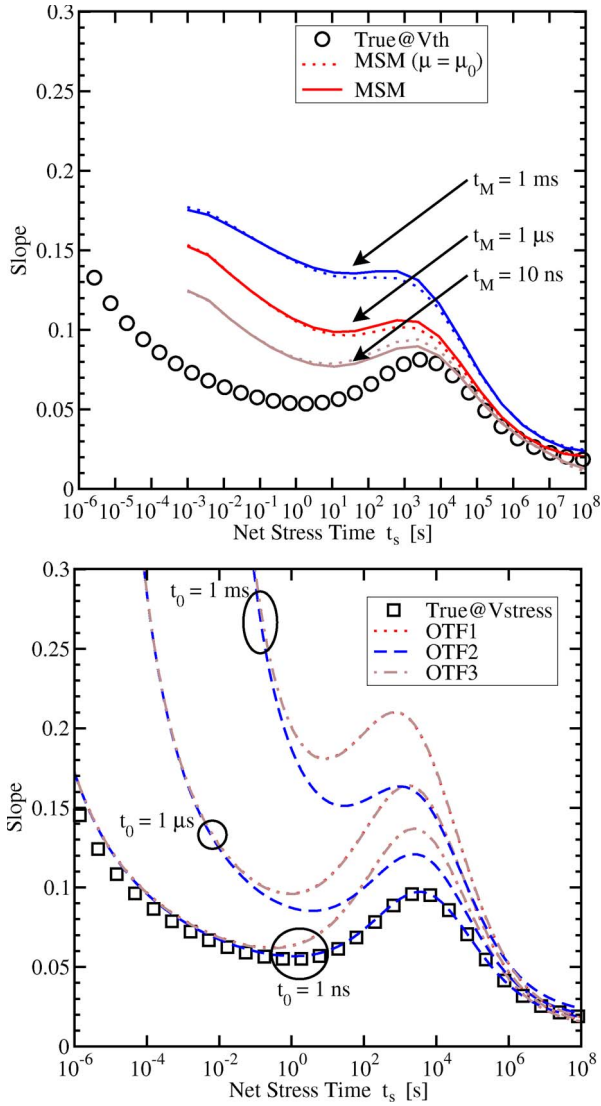


Fig. 11. Influence of the measurement method on the observed slope of an effective power-law time dependence. Whereas the (top) MSM methods depend primarily on the measurement delay t_M , OTF1 and OTF3 are contaminated by t_0 , the mobility variation, and compact modeling errors ($\theta = \text{const}$, etc.). Only an extremely fast OTF2 method with a hypothetical initial delay of 1 ns is able to reproduce the expected values. The fastest hypothetical MSM method used a measurement delay of 10 ns; smaller delays resulted in distortions due to interface state dynamics.

Fig. 13 shows that, due to the early increase in ΔN_{it} , a distortion in the extracted ΔV_{th} is obtained also at earlier times.

V. CONCLUSION

We have analyzed the most commonly used NBTI measurement techniques. Our analysis is based on a rigorous numerical solution of the semiconductor device equations, augmented by suitable models for interface and oxide charges. As expected, it is confirmed that MSM methods are distorted by unavoidable delay. Interestingly, MSM methods also contain a small but temperature-dependent spurious threshold-voltage shift caused by mobility degradation but appear to be undistorted by the interface state dynamics. OTF methods, on the other hand, suffer from the problem of the initial measurement and can

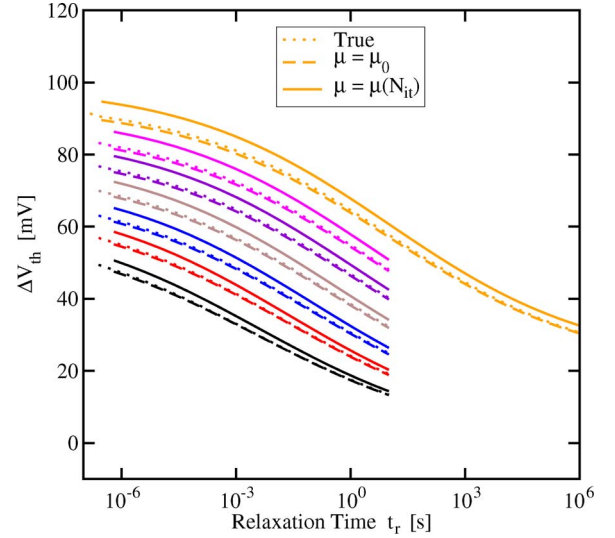


Fig. 12. The same as Fig. 5, but assuming that only interface states determine the degradation, $N_{it} = R + P$. Now, the 10% mobility variation introduces a constant 5% error, rather than a 4.5-mV offset.

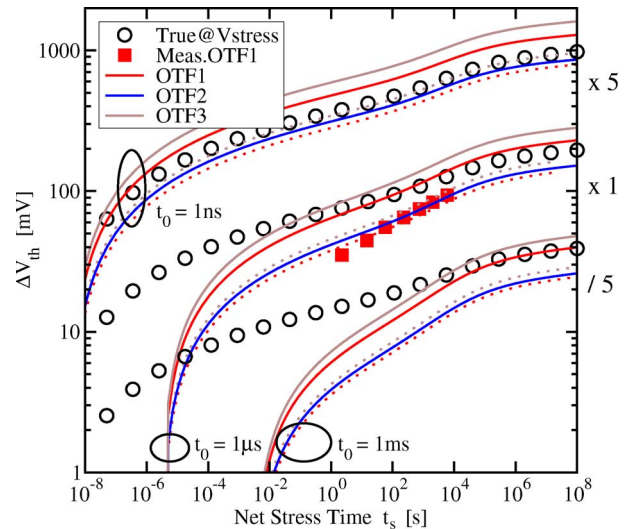


Fig. 13. The same as Fig. 9, but assuming that only interface states determine the degradation, $N_{it} = R + P$. Because the mobility model assumes a mobility degradation as a function of N_{it} , a distortion is also obtained at earlier times.

be polluted by rather complex mobility changes. Furthermore, all OTF methods rely on simple compact models for the back-extrapolation of the threshold-voltage shift at stress-level, which results in difficult-to-quantify inaccuracies.

Overall, it could be confirmed that all measurement techniques used to determine the degradation after bias temperature stress are prone to errors. Nevertheless, in spite of the inherent delay, the smaller systematic errors in the conventional MSM technique allow one to accurately monitor the time evolution of the recovery, which can then be used for the formulation of physics-based models.

ACKNOWLEDGMENT

The authors would like to thank H. Reisinger, V. Huard, S. Mahapatra, and N. Mielke for the fruitful discussions.

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